OUT OF BAND AND ICI REDUCTION TECHNIQUE

By : Hamid Aminoroaya • There is a substantial need for more frequency bandwidth and the efficient and flexible use of existing bands.



OFDM system

$$s(t) = \sum_{i=-\infty}^{+\infty} \sum_{k=1}^{N_{SC}} c_{ki} e^{j2\pi f_k(t-iT_s)} f(t-iT_s)$$



Out of band (OOB) component • $S(x) = \sum_{n=-N/2}^{N/2-1} dn \cdot \operatorname{sinc} (x - xn)$, $x = (f - f_o) T_o$

• The sidelobe power of this sum signal only decays with $1/(x^2 N)$ resulting in a high out-of-band radiation.



OOB reduction technique

- Windowing
- Guard band
- Cancellation Carrier (CC)
- Subcarrier Weighting (SW)
- Multiple Choice Sequences (MCS)
- Constellation Expansion (CE)
- Additive Signal Method (AS)
- Combined methods

Windowing

• The signal *r*(*k*) is windowed in the time domain by a window function *w*(*k*) :

 $r^{(k)} = r(k) w(k)$

- making the PDS of an OFDM modulated carrier go down more rapidly by windowing the transmit signal of the OFDM symbols.
- A commonly used window type is the raised cosine window.
- drawback of this method is that windowing expands the signal in time domain and intersymbol interference (ISI) is introduced.

Windowing

• Even at very high rolloff factors, the achievable interference reduction is only about 6dB.



n = number of adjacent sub-band

Guard band



The drawback of this method is the less effective use of the available bandwidth.

Guard band



b = number of deactivated adjacent subcarriers

deactivation of the first adjacent subcarrier (b = 1)*delivers the largest* benefit. The additional deactivation of more subcarriers $(b \ge 2)$ only provides a minor further improvement.

Cancellation Carrier (CC)

- Each CC is multiplied by a *complex weighting factor gm*
- the transmit symbol is modulated on N + M subcarriers



Cancellation Carrier (CC)

$$\min_{\mathbf{g}} \left\| \mathbf{s} + \sum_{m=1}^{M} g_m \cdot \mathbf{c}_m \right\|^2 \quad \text{subject to} \quad \|\mathbf{g}\|^2 \le \alpha$$

- The constraint limits the power of the CCs to *α* in order not to spend too much Tx power on the CCs.
- with only two CCs at each side of the used spectrum the out-of-band radiation can be reduced by more than 20 dB.
- The price is
 Loss in BER performance
 Increased computational complexity.

Subcarrier Weighting (SW)

 multiplication of each symbol *dn* with a real valued weighting factor *gn*.

Pulse shaping



Subcarrier Weighting (SW)

- solving the optimization problem with two constraints:
 - Keeps the transmission power the same as in the case without weighting i.e. ||⁻d ||² = ||d||².
 - elements of g are between pre-defined limits,
 i.e., gmin ≤ gn ≤ gmax

•
$$\rho = gmax/gmin/\longrightarrow$$
 bit-error rate (BER)/
• OOB/

Multiple Choice Sequences (MCS)

- The principle is to map original transmission sequence into another transmission sequence, which has lower sidelobes.
- algorithms to generate an MCS set :
 - Symbol constellation approach
 - Interleaving approach
 - Phase approach



Constellation Expansion (CE)

• Exploiting the fact that different sequences have different sidelobe power levels



Additive Signal Method (AS)

- optimization problem with two constraints:
 - Transmission power the same as in the original sequence
 - elements of **a** are between pre-defined limits , i.e., $||an|| \le R$



There is a trade-off between the additional sidelobe suppression obtained by enlarging the radius R and the increased loss in SNR performance.

Combined methods

- CC + CE
- CC + windowing
- MCS + CC
- MCS + SW
- SW + guard band

Inter carrier interference (ICI)

- OFDM is very sensitive to frequency errors, caused by :
 - Carrier frequency mismatch between the transmitter and receiver
 - The Doppler shift.
- Leads to an orthogonality-loss between carriers and intercarrier interference (ICI) will occur.

Inter carrier interference (ICI)

Received signal

$$y(t) = \exp j(2\pi\Delta ft + \theta_0) \sum_{k=0}^{N-1} b_{k,i} q\left(t - \frac{kT}{N}\right)$$

• After FFT

$$Y(k) = X(k)S(0) + \sum_{l=0, l \neq k}^{N-1} X(l)S(l-k) + n_k,$$

carrier-to-interference power ratio (CIR)

$$\text{CIR} = \frac{|S(k)|^2}{\sum\limits_{l=0,\,l\neq k}^{N-1} |S(l-k)|^2} = \frac{|S(0)|^2}{\sum\limits_{l=1}^{N-1} |S(l)|^2}.$$

ICI reduction methods

- ICI self-cancellation
- pulse shaping
- select mapping
- frequency-domain equalization
- time-domain windowing

ICI self-cancellation

- the difference between S(l-k) and S(l-k+1) is very small.
- If a data pair is modulated onto two adjacent subcarriers (a,-a), then the ICI signals generated by the subcarriers will cancel themselves.

ICI self-cancellation

ICI self-cancellation scheme can also be extended to group of L subcarriers.



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pulse shaping

$$x(t) = \exp\{j2\pi f_c t\} \sum_{k=0}^{N-1} s_k p(t) \exp\{j2\pi f_k t\}$$

• Rectangular pulse (REC)

$$P_{REC}(f) = \sin c(fT)$$

- Raised cosine pulse (RC) $P_{RC}(f) = \sin c(fT) \frac{\cos(\pi o fT)}{1 - (2 o fT)^2},$
- Better than raised cosine pulse (BTRC) $P_{BTRC}(f) = \sin c(fT) \frac{2\beta fT \sin(\pi \alpha fT) + 2\cos(\pi \alpha fT) - 1}{1 + (\beta fT)^2}$
- Sinc power pulse (SP) $P_{SP}(f) = \sin c^n (fT)$
- Improved sinc power pulse (ISP)

 $P_{ISP}(f) = \exp\{-a(fT)^2\}\sin^n(fT),$

pulse shaping



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